## DATA SHEET

## TDA8356

DC-coupled vertical deflection circuit

File under Integrated Circuits, IC02

## FEATURES

- Few external components
- Highly efficient fully DC-coupled vertical output bridge circuit
- Vertical flyback switch
- Guard circuit
- Protection against:
- short-circuit of the output pins (7 and 4)
- short-circuit of the output pins to $\mathrm{V}_{\mathrm{P}}$
- Temperature (thermal) protection
- High EMC immunity because of common mode inputs
- A guard signal in zoom mode.


## GENERAL DESCRIPTION

The TDA8356 is a power circuit for use in $90^{\circ}$ and $110^{\circ}$ colour deflection systems for field frequencies of 50 to 120 Hz . The circuit provides a DC driven vertical deflection output circuit, operating as a highly efficient class $G$ system.

## QUICK REFERENCE DATA

| SYMBOL | PARAMETER | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DC supply |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{P}}$ | supply voltage | 9 | 4.5 | 25 | V |
| $\mathrm{I}_{\mathrm{q}}$ | quiescent supply current | - | 30 | - | mA |
| Vertical circuit |  |  |  |  |  |
| $\mathrm{l}_{\mathrm{O}(\mathrm{p}-\mathrm{p})}$ | output current (peak-to-peak value) | - | - | 2 | A |
| $I_{\text {diff( } p \text {-p) }}$ | differential input current (peak-to-peak value) | - | 600 | - | $\mu \mathrm{A}$ |
| $V_{\text {diff( }}(p-p)$ | differential input voltage (peak-to-peak value) | - | 1.5 | 1.8 | V |
| Flyback switch |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{M}}$ | peak output current | - | - | $\pm 1$ | A |
| $\mathrm{V}_{\mathrm{FB}}$ | flyback supply voltage | - | - | 50 | V |
| Thermal data (in accordance with IEC 747-1) |  |  |  |  |  |
| $\mathrm{T}_{\text {stg }}$ | storage temperature | -55 | - | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {amb }}$ | operating ambient temperature | -25 | - | +75 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{vj}}$ | virtual junction temperature | - | - | 150 | ${ }^{\circ} \mathrm{C}$ |

## ORDERING INFORMATION

| TYPE NUMBER | PACKAGE |  |  |
| :---: | :---: | :--- | :---: |
|  | NAME | DESCRIPTION | VERSION |
| TDA8356 | SIL9P | plastic single-in-line power package; 9 leads | SOT131-2 |

## BLOCK DIAGRAM



Fig. 1 Block diagram.

## PINNING

| SYMBOL | PIN | DESCRIPTION |
| :--- | :---: | :--- |
| $I_{\text {drive(pos) }}$ | 1 | input power-stage (positive); <br> includes $\mathrm{I}_{\text {I(sb) }}$ signal bias |
| $\mathrm{I}_{\text {drive(neg) }}$ | 2 | input power-stage (negative); <br> includes $\mathrm{I}_{\text {I(sb) }}$ signal bias |
| $\mathrm{V}_{\mathrm{P}}$ | 3 | operating supply voltage |
| $\mathrm{V}_{\mathrm{O}(\mathrm{B})}$ | 4 | output voltage B |
| GND | 5 | ground |
| $\mathrm{V}_{\mathrm{FB}}$ | 6 | input flyback supply voltage |
| $\mathrm{V}_{\mathrm{O}(\mathrm{A})}$ | 7 | output voltage A |
| $\mathrm{V}_{\mathrm{O}(\text { guard }}$ | 8 | guard output voltage |
| $\mathrm{V}_{\text {Ifb) }}$ | 9 | input feedback voltage |



Metal block connected to substrate pin 5.
Metal on back.
Fig. 2 Pin configuration.

## FUNCTIONAL DESCRIPTION

The vertical driver circuit is a bridge configuration. The deflection coil is connected between the output amplifiers, which are driven in phase opposition. An external resistor $\left(R_{M}\right)$ connected in series with the deflection coil provides internal feedback information. The differential input circuit is voltage driven. The input circuit has been adapted to enable it to be used with the TDA9150, TDA9151B, TDA9160A, TDA9162, TDA8366 and TDA8376 which deliver symmetrical current signals. An external resistor ( $\mathrm{R}_{\mathrm{CON}}$ ) connected between the differential input determines the output current through the deflection coil. The relationship between the differential input current and the output current is defined by: $I_{\text {diff }} \times R_{\text {CON }}=I_{\text {coil }} \times R_{M}$. The output current is adjustable from $0.5 \mathrm{~A}(\mathrm{p}-\mathrm{p})$ to $2 \mathrm{~A}(p-p)$ by varying $R_{M}$. The maximum input differential voltage is 1.8 V . In the application it is recommended that $\mathrm{V}_{\text {diff }}=1.5 \mathrm{~V}$ (typ). This is recommended because of the spread of input current and the spread in the value of $\mathrm{R}_{\mathrm{CON}}$.
The flyback voltage is determined by an additional supply voltage $\mathrm{V}_{\mathrm{FB}}$. The principle of operating with two supply voltages (class $G$ ) makes it possible to fix the supply voltage $\mathrm{V}_{\mathrm{P}}$ optimum for the scan voltage and the second supply voltage $\mathrm{V}_{\mathrm{FB}}$ optimum for the flyback voltage. Using this method, very high efficiency is achieved.

The supply voltage $\mathrm{V}_{\mathrm{FB}}$ is almost totally available as flyback voltage across the coil, this being possible due to the absence of a decoupling capacitor (not necessary, due to the bridge configuration). The output circuit is fully protected against the following:

- thermal protection
- short-circuit protection of the output pins (pins 4 and 7 )
- short-circuit of the output pins to $\mathrm{V}_{\mathrm{P}}$.

A guard circuit $\mathrm{V}_{\mathrm{O} \text { (guard) }}$ is provided. The guard circuit is activated at the following conditions:

- during flyback
- during short-circuit of the coil and during short-circuit of the output pins (pins 4 and 7 ) to $\mathrm{V}_{\mathrm{P}}$ or ground
- during open loop
- when the thermal protection is activated.

This signal can be used for blanking the picture tube screen.

## LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

| SYMBOL | PARAMETER | CONDITIONS | MIN. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DC supply |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{P}}$ | supply voltage | non-operating | - | 40 | V |
|  |  |  | - | 25 | V |
| $\mathrm{V}_{\mathrm{FB}}$ | flyback supply voltage |  | - | 50 | V |
| Vertical circuit |  |  |  |  |  |
| $\mathrm{I}_{(\text {(p-p) }}$ | output current (peak-to-peak value) | note 1 | - | 2 | A |
| $\mathrm{V}_{\mathrm{O}(\mathrm{A})}$ | output voltage (pin 7) |  | - | 52 | V |
| Flyback switch |  |  |  |  |  |
| $\mathrm{I}_{\mathrm{M}}$ | peak output current |  | - | $\pm 1.5$ | A |
| Thermal data (in accordance with IEC 747-1) |  |  |  |  |  |
| $\mathrm{T}_{\text {stg }}$ | storage temperature |  | -55 | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {amb }}$ | operating ambient temperature |  | -25 | +75 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{vj}}$ | virtual junction temperature |  | - | 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{R}_{\text {th } \mathrm{vj-c}}$ | resistance $\mathrm{v}_{\mathrm{j}}$-case |  | - | 4 | K/W |
| $\mathrm{R}_{\text {th } \mathrm{j} j-\mathrm{a}}$ | resistance $v_{j}$-ambient in free air |  | - | 40 | K/W |
| $\mathrm{t}_{\mathrm{sc}}$ | short-circuiting time | note 2 | - | 1 | hr |

## Notes

1. I I maximum determined by current protection.
2. Up to $V_{P}=18 \mathrm{~V}$.

## CHARACTERISTICS

$V_{P}=14.5 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{FB}}=45 \mathrm{~V} ; \mathrm{f}_{\mathrm{i}}=50 \mathrm{~Hz} ; \mathrm{I}_{(\mathrm{sb})}=400 \mu \mathrm{~A}$; measured in test circuit of Fig.3; unless otherwise specified.

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DC supply |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{P}}$ | operating supply voltage |  | 9.0 | 4.5 | 25 | V |
| $\mathrm{V}_{\mathrm{FB}}$ | flyback supply voltage |  | $\mathrm{V}_{\mathrm{P}}$ | - | 50 | V |
| $\mathrm{I}_{\mathrm{P}}$ | supply current | no signal; no load | - | 30 | 55 | mA |
| Vertical circuit |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{O}}$ | output voltage swing (scan) | $\begin{array}{\|l} \hline \mathrm{I}_{\text {diff }}=0.6 \mathrm{~mA}(p-p) ; \\ \mathrm{V}_{\text {diff }}=1.8 \mathrm{~V}(\mathrm{p}-\mathrm{p}) ; \\ \mathrm{I}_{\mathrm{O}}=2 \mathrm{~A}(\mathrm{p-p}) \\ \hline \end{array}$ | 13.2 | - | - | V |
| LE | linearity error | $\mathrm{I}_{\mathrm{O}}=2 \mathrm{~A}(\mathrm{p}-\mathrm{p})$; note 1 | - | 1 | 4 | \% |
|  |  | $\mathrm{l}_{\mathrm{O}}=50 \mathrm{~mA}(\mathrm{p}-\mathrm{p})$; note 1 | - | 1 | 4 | \% |
| $\mathrm{V}_{\mathrm{O}}$ | output voltage swing (flyback) $V_{O(A)}-V_{O(B)}$ | $\begin{aligned} & \mathrm{I}_{\text {diff }}=0.3 \mathrm{~mA} ; \\ & \mathrm{I}_{\mathrm{O}}=1 \mathrm{~A}(\mathrm{M}) \end{aligned}$ | - | 40 | - | V |
| $\mathrm{V}_{\mathrm{DF}}$ | forward voltage of the internal efficiency diode ( $\mathrm{V}_{\mathrm{O}(\mathrm{A})}$ - $\mathrm{V}_{\mathrm{FB}}$ ) | $\begin{aligned} & \mathrm{I}_{\mathrm{O}}=-1 \mathrm{~A}(\mathrm{M}) ; \\ & \mathrm{I}_{\text {diff }}=0.3 \mathrm{~mA} \end{aligned}$ | - | - | 1.5 | V |
| \| ${ }_{\text {os }}$ \| | output offset current | $\begin{aligned} & I_{\text {diff }}=0 ; \\ & I_{(\mathrm{sb})}=50 \text { to } 500 \mu \mathrm{~A} \end{aligned}$ | - | - | 40 | mA |
| $\left\|\mathrm{V}_{\text {os }}\right\|$ | offset voltage at the input of the feedback amplifier $\left(\mathrm{V}_{\mathrm{l} \text { (fb) }}-\mathrm{V}_{\mathrm{O}(\mathrm{B})}\right)$ | $\begin{aligned} & \mathrm{I}_{\text {diff }}=0 ; \\ & \mathrm{I}_{(\text {(sb })}=50 \text { to } 500 \mu \mathrm{~A} \\ & \hline \end{aligned}$ | - | - | 24 | mV |
| $\Delta \mathrm{V}_{\text {os }} \mathrm{T}$ | output offset voltage as a function of temperature | $\mathrm{I}_{\text {diff }}=0$ | - | - | 72 | $\mu \mathrm{V} / \mathrm{K}$ |
| $\mathrm{V}_{\mathrm{O}(\mathrm{A})}$ | DC output voltage | $\mathrm{I}_{\text {diff }}=0 ;$ note 2 | - | 6.5 | - | V |
| $\mathrm{G}_{\mathrm{vo}}$ | open-loop voltage gain ( $\left.\mathrm{V}_{7-4} / \mathrm{V}_{1-2}\right)$ | notes 3 and 4 | - | 80 | - | dB |
|  | open loop voltage gain $\left(V_{7-4} / V_{9-4} ; V_{1-2}=0\right)$ | note 3 | - | 80 | - | dB |
| $\mathrm{V}_{\mathrm{R}}$ | voltage ratio $\mathrm{V}_{1-2} / \mathrm{V}_{9-4}$ |  | - | 0 | - | dB |
| $\mathrm{f}_{\text {res }}$ | frequency response (-3 dB) | open loop; note 5 | - | 40 | - | Hz |
| $\mathrm{G}_{1}$ | current gain ( $\mathrm{l}_{\mathrm{o}} / \mathrm{ldifif}^{\text {) }}$ |  | - | 5000 | $-$ |  |
| $\Delta \mathrm{G}_{\mathrm{c}} \mathrm{T}$ | current gain drift as a function of temperature |  | - | - | $10^{-4}$ | K |
| $I_{1(\text { sb) }}$ | signal bias current |  | 50 | 400 | 500 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {FB }}$ | flyback supply current | during scan | - | - | 100 | $\mu \mathrm{A}$ |
| PSRR | power supply ripple rejection | note 6 | - | 80 | - | dB |
| $\mathrm{V}_{1(\mathrm{DC})}$ | DC input voltage |  | - | 2.7 | - | V |
| $\mathrm{V}_{\text {I(CM) }}$ | common mode input voltage | $\mathrm{I}_{(\text {(sb) }}=0$ | 0 | - | 1.6 | V |
| $\mathrm{I}_{\text {bias }}$ | input bias current | $\mathrm{I}_{(\text {(sb) })}=0$ | - | 0.1 | 0.5 | $\mu \mathrm{A}$ |
| $\mathrm{IO}_{\mathrm{O}} \mathrm{CM}$ ) | common mode output current | $\begin{aligned} & \Delta \mathrm{I}_{(\text {sb })}=300 \mu \mathrm{~A}(\mathrm{p}-\mathrm{p}) ; \\ & \mathrm{f}_{\mathrm{i}}=50 \mathrm{~Hz} ; \mathrm{I}_{\text {diff }}=0 \end{aligned}$ | - | 0.2 | - | mA |


| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Guard circuit |  |  |  |  |  |  |
| $\mathrm{l}_{0}$ | output current | not active; $\mathrm{V}_{\mathrm{O} \text { (guard) }}=0 \mathrm{~V}$ | - | - | 50 | $\mu \mathrm{A}$ |
|  |  | active; $\mathrm{V}_{\mathrm{O}(\text { guard })}=4.5 \mathrm{~V}$ | 1 | - | 2.5 | mA |
| $\mathrm{V}_{\text {O(guard) }}$ | output voltage on pin 8 | $\mathrm{I}_{\mathrm{O}}=100 \mu \mathrm{~A}$ | - | - | 5.5 | V |
|  | allowable voltage on pin 8 | maximum leakage current $=10 \mu \mathrm{~A}$; | - | - | 40 | V |

## Notes

1. The linearity error is measured without S-correction and based on the same measurement principle as performed on the screen. The measuring method is as follows:
Divide the output signal $\mathrm{I}_{4}-\mathrm{I}_{7}\left(\mathrm{~V}_{\mathrm{RM}}\right)$ into 22 equal parts ranging from 1 to 22 inclusive. Measure the value of two succeeding parts called one block starting with part 2 and 3 (block 1) and ending with part 20 and 21 (block 10). Thus part 1 and 22 are unused. The equations for linearity error for adjacent blocks (LEAB) and not adjacent blocks (NAB) are given below
LEAB $=\frac{\mathrm{a}_{\mathrm{k}}-\mathrm{a}_{(\mathrm{k}+1)}}{\mathrm{a}_{\mathrm{avg}}} ; N A B=\frac{\mathrm{a}_{\text {max }}-\mathrm{a}_{\text {min }}}{\mathrm{a}_{\mathrm{avg}}}$
2. Related to $\mathrm{V}_{\mathrm{P}}$.
3. V values within formulae, relate to voltages at or between relative pin numbers, i.e. $\mathrm{V}_{7-4} / \mathrm{V}_{1-2}=$ voltage value across pins 7 and 4 divided by voltage value across pins 1 and 2.
4. $\mathrm{V}_{9-4} \mathrm{AC}$ short-circuited.
5. Frequency response $\mathrm{V}_{7-4} / \mathrm{V}_{9-4}$ is equal to frequency response $\mathrm{V}_{7-4} / \mathrm{V}_{1-2}$.
6. At $\mathrm{V}_{\text {(ripple) }}=500 \mathrm{mV}$ eff; measured across $\mathrm{R}_{\mathrm{M}} ; \mathrm{f}_{\mathrm{i}}=50 \mathrm{~Hz}$.


Fig. 3 Test diagram.


Fig. 4 Input currents.

## APPLICATION INFORMATION


$\mathrm{V}_{\mathrm{P}}=13.5 \mathrm{~V} ; \mathrm{I}_{\mathrm{O}(\mathrm{p}-\mathrm{p})}=1.87 \mathrm{~A} ; \mathrm{I}_{(\mathrm{sb})}=400 \mu \mathrm{~A} ; \mathrm{I}_{\text {diff(p-p) }}=500 \mu \mathrm{~A} ; \mathrm{V}_{\mathrm{FB}}=42 \mathrm{~V} ; \mathrm{t}_{\mathrm{FB}}=0.6 \mathrm{~ms}$.
Fig. 5 Application diagram.

## PACKAGE OUTLINE



Dimensions in mm.
Fig. 6 Plastic single-in-line power package; 9 leads (SIL9P; SOT131-2).

## SOLDERING

## Plastic single in-line packages

BY DIP OR WAVE
The maximum permissible temperature of the solder is $260^{\circ} \mathrm{C}$; this temperature must not be in contact with the joint for more than 5 s . The total contact time of successive solder waves must not exceed 5 s .

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the
specified storage maximum. If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

## Repairing soldered joints

Apply the soldering iron below the seating plane (or not more than 2 mm above it. If its temperature is below $300^{\circ} \mathrm{C}$, it must not be in contact for more than 10 s ; if between 300 and $400^{\circ} \mathrm{C}$, for not more than 5 s .

## DEFINITIONS

| Data sheet status |  |
| :--- | :--- |
| Objective specification | This data sheet contains target or goal specifications for product development. |
| Preliminary specification | This data sheet contains preliminary data; supplementary data may be published later. |
| Product specification | This data sheet contains final product specifications. |
| Limiting values | Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or <br> more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation <br> of the device at these or at any other conditions above those given in the Characteristics sections of this specification <br> is not implied. Exposure to limiting values for extended periods may affect device reliability. |
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